

# Digital Communication Systems Engineering with Software-Defined Radio

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## Lecture 12

# Orthonormal Basis Functions

- ▶ Recall from Lecture 10 that  $\{\phi_j(t)\}$  is an orthonormal set of functions on the time interval  $[0, T]$  such that:

$$\int_0^T \phi_i(t)\phi_j(t)dt = \begin{cases} 1 & i = j \\ 0 & \text{otherwise} \end{cases}$$

- ▶ Furthermore, it is possible to represent a signal waveform  $s_i(t)$  as the weighted sum of these orthonormal basis functions, i.e.:

$$s_i(t) = \sum_{k=1}^N s_{ik}\phi_k(t) \quad (1)$$

- ▶ How do we determine the set of orthonormal basis functions  $\{\phi_j(t)\}$ ?
  - ▶ Use *Gram-Schmidt Orthogonalization Procedure*

# Orthonormal Basis Functions

- ▶ A complete orthonormal set of basis functions is needed for a set of  $M$  energy signals denoted by  $s_1(t), \dots, s_M(t)$
- ▶ Choose  $s_1(t)$  and normalize it:

$$\phi_1(t) = \frac{s_1(t)}{\sqrt{E_{s_1}}}$$

where  $E_{s_1}$  is the energy of the signal  $s_1(t)$

- ▶ We can express  $s_1(t)$  as:

$$s_1(t) = \sqrt{E_{s_1}} \phi_1(t) = s_{11} \phi_1(t)$$

where the coefficient  $s_{11} = \sqrt{E_{s_1}}$  and  $\phi_1(t)$  has unit energy as required

## Setting Up $\phi_2(t)$

- ▶ Now using the signal  $s_2(t)$ , let us define the  $s_{21}$  coefficient as:

$$s_{21} = \int_0^T s_2(t)\phi_1(t)dt$$

- ▶ To help in getting the basis function  $\phi_2(t)$ , we define the intermediate function:

$$g_2(t) = s_2(t) - s_{21}\phi_1(t)$$

which is orthogonal to  $\phi_1(t)$  over the interval  $0 \leq t \leq T$

# Creating the Second Basis Function

- ▶ The second basis function can be defined as:

$$\phi_2(t) = \frac{g_2(t)}{\sqrt{\int_0^T g_2^2(t) dt}} \quad (2)$$

which can be expanded to:

$$\phi_2(t) = \frac{s_2(t) - s_{21}\phi_1(t)}{\sqrt{E_{s_2} - s_{21}^2}} \quad (3)$$

where  $E_{s_2}$  is the energy of the signal  $s_2(t)$ .

- ▶ It is clear that the basis function  $\phi_2(t)$  satisfies:

$$\int_0^T \phi_2^2(t) dt = 1 \quad \text{and} \quad \int_0^T \phi_1(t)\phi_2(t) dt = 0$$

# General Expressions

- In general, we can define the following functions that can be employed in an iterative procedure for generating a set of orthonormal basis functions:

$$g_i(t) = s_i(t) - \sum_{j=1}^{i-1} s_{ij} \phi_j(t)$$

$$s_{ij} = \int_0^T s_i(t) \phi_j(t) dt, \quad j = 1, 2, \dots, i-1$$

$$\phi_i(t) = \frac{g_i(t)}{\sqrt{\int_0^T g_i^2(t) dt}}, \quad i = 1, 2, \dots, N$$

# Problem Statement

- ▶ Carry out the Gram-Schmidt Orthogonalization procedure of the signals shown in the next figure in the order  $s_3(t)$ ,  $s_1(t)$ ,  $s_4(t)$ ,  $s_2(t)$  and thus obtain a set of orthonormal functions  $\{\phi_m(t)\}$
- ▶ Then, determine the vector representation of the signals  $\{s_n(t)\}$  by using the orthonormal functions  $\{\phi_m(t)\}$

# Signal Waveforms

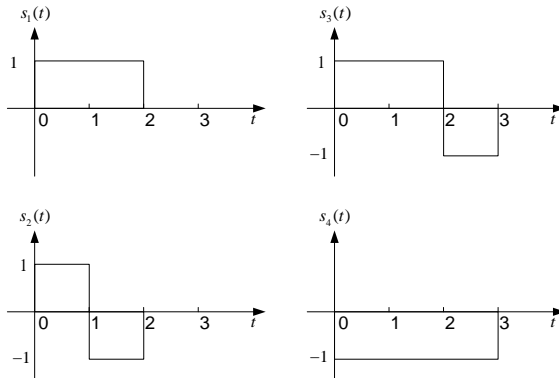


Figure : Example Signal Waveforms.

# Determining Orthonormal Basis Functions – Part I

- ▶ For  $s_3(t)$ :

$$\phi_1(t) = \frac{s_3(t)}{\sqrt{E_{s_3}}} = \frac{s_3(t)}{\sqrt{3}}$$

- ▶ For  $s_1(t)$ :

$$g_2(t) = s_1(t) - s_{12}\phi_1(t) = s_1(t) - \frac{2}{3}s_3(t) = \begin{cases} 1/3, & 0 \leq t < 2 \\ 2/3, & 2 \leq t < 3 \\ 0, & t \geq 3 \end{cases}$$

$$\therefore \phi_2(t) = \frac{g_2(t)}{\sqrt{\int_0^T g_2^2(t) dt}} = \begin{cases} 1/\sqrt{6}, & 0 \leq t < 2 \\ 2/\sqrt{6}, & 2 \leq t < 3 \\ 0, & t \geq 3 \end{cases}$$

# Determining Orthonormal Basis Functions – Part II

- For  $s_4(t)$ :

$$g_3(t) = s_4(t) - \sum_{j=1}^2 s_{4j} \phi_j(t) = 0$$
$$\therefore \phi_3(t) = 0$$

- For  $s_4(t)$ :

$$g_4(t) = s_2(t) - \sum_{j=1}^3 s_{2j} \phi_j(t) = 0$$
$$\therefore \phi_4(t) = \frac{g_4(t)}{\sqrt{\int_0^T g_4^2(t) dt}} = \frac{s_2(t)}{\sqrt{2}}$$

# Vector Representations

- ▶ Using the orthonormal basis functions  $\{\phi_1(t), \phi_2(t), \phi_4(t)\}$ , we can express the four waveforms as:
  - ▶  $\mathbf{s}_1 = (2/\sqrt{3}, \sqrt{6}/3, 0)$
  - ▶  $\mathbf{s}_2 = (0, 0, \sqrt{2})$
  - ▶  $\mathbf{s}_3 = (\sqrt{3}, 0, 0)$
  - ▶  $\mathbf{s}_4 = (-1/\sqrt{3}, -4/\sqrt{6}, 0)$